

1 Maximizing Archimedes spiral packing density 2 area

3 **DHAIFALLAH ALMUTAIRI,^{1,2,*} NAIF ALSHAMRANI,^{1,2} ANDREW
4 INGRAM,¹ ANDREW GRIECO,¹ AND YESHAIAHU FAINMAN¹**

5 ¹*Department of Electrical & Computer Engineering, University of California, San Diego, 9500 Gilman
6 Dr., La Jolla, California 92093, USA.*

7 ²*King Abdulaziz City for Science and Technology (KACST), P.O. Box 6086, Riyadh 11442, Saudi Arabia.
8 *daalmuta@eng.ucsd.edu*

9 **Abstract:** In this paper, we experimentally demonstrate a broadband Archimedes spiral delay
10 line with high packing density on a silicon photonic platform. This high density is achieved by
11 optimizing the gap between the adjacent waveguides (down to sub-micron scale) in the spiral
12 configuration. However, care must be taken to avoid evanescent coupling, the presence of
13 which will cause the spiral to behave as a novel type of distributed spiral resonator. To this end,
14 an analytical model of the resonance phenomenon was developed for a simple spiral. Moreover,
15 it is demonstrated that this distributed spiral resonator effect can be minimized by ensuring that
16 adjacent waveguides in the spiral configuration have different propagation constants (β).
17 Experimental validations were accomplished by fabricating and testing multiple spiral
18 waveguides with varying lengths (i.e., 0.4, 0.8, and 1.4 mm) and separation gaps (i.e., 300 and
19 150 nm). Finally, a Linear Density Figure of Merit (LDFM) is introduced to evaluate the
20 packing efficiency of various spiral designs in the literature. In this work, the optimum
21 experimental design with mitigated resonance had a length of 1.4mm and occupied an area of
22 60x60 μ m, corresponding to an LDFM of 388mm⁻¹.

23 © 2022 Optica Publishing Group under the terms of the [Optica Publishing Group Open Access Publishing](#)
24 [Agreement](#)

25 1. Introduction

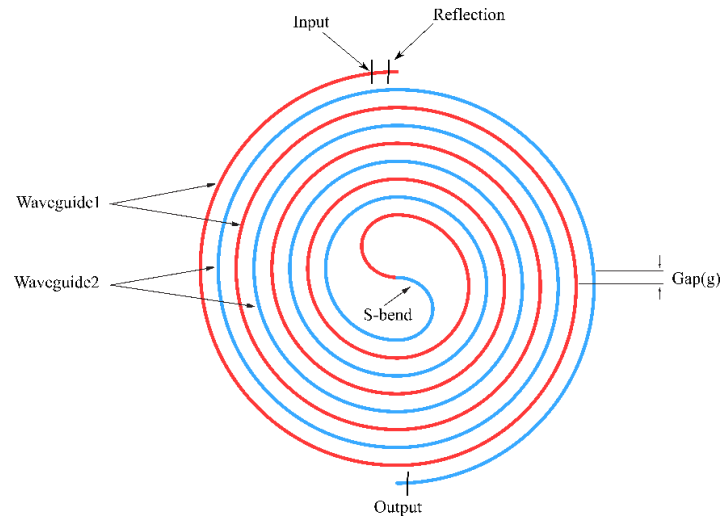
26 The Optical Delay Line (ODL) is an essential building block in numerous applications,
27 including optical communications [1], microwave signal processing [2,3], optical gyroscopes
28 [4], optical coherence tomography (OCT) [5], phased arrays [6], quantum computing [7],
29 reservoir computing [8] and spectrometers [9]. Compact implementations of ODLs have been
30 demonstrated employing several resonant designs and structures such as Bragg gratings [10],
31 photonic crystals [11,12], coupled resonator structures [13], and stimulated Brillouin scattering
32 photonic filters [14,15]. However, designs that use resonance phenomena to reduce the
33 footprint of the ODL are limited in operation to discrete narrow optical frequencies in a limited
34 spectral range. The operating bandwidth can be enhanced by cascading several resonators to
35 increase the overall device operating bandwidth; however, the device's footprint increases
36 commensurately [16]. Many applications, however, require true ODLs (TODL) that can work
37 in a large optical frequency spectral band that cannot be supported with these resonance-based
38 approaches. An alternative approach to constructing a TODL is to employ a long waveguide in
39 a serpentine [17] or spiral [18] configuration to reduce the area footprint. In such a device, the
40 accumulated time delay can be estimated through the following relation: $= \frac{n_g L}{c}$ Where n_g , L , and
41 c are the group refractive index, the waveguide length, and the speed of light in a vacuum.

42 This manuscript proposes and demonstrates a TODL using an enhanced Archimedes spiral
43 configuration to enable high packing density and small footprint [16,19]. Crucially, the
44 Archimedes spiral in this work consists of clockwise and counterclockwise waveguides with
45 differing propagation constants (β). A tapered S-shape bend waveguide to adiabatically connect
46 these two spiral waveguides [16,19]. The mismatch minimizes the evanescent cross-coupling

47 between adjacent waveguides and therefore allows them to be packed much more closely than
48 in conventional designs. This approach can reduce the gap between adjacent waveguides down
49 to the submicron scale while maintaining negligibly small evanescent coupling. Validation is
50 performed analytically, numerically, and experimentally.

51 2. Design, simulation, and modeling of TODL

52 Theoretically, when two waveguides are placed close to each other, they become evanescently
53 coupled, which results in crosstalk [20]. According to the coupled-mode theory, such crosstalk
54 will be at its highest when both waveguides are designed to have identical propagation
55 constants(β). This crosstalk can be mitigated by introducing a mismatch between the adjacent
56 waveguide propagation constants, such that the larger the mismatch, the smaller the cross talk.
57 This effect is significant because conventional spiral designs use adjacent waveguides with
58 matched propagation constants, which strongly limits how close the waveguides can be packed
59 together. The two adjacent waveguides should exhibit different propagation constants to
60 overcome this limitation and increase the packing density. The Archimedean spiral delay line
61 of such a structure is schematically shown in Fig. 1. The bending radius of the S-shape bend is
62 kept as small as possible while avoiding the introduction of bending radiative loss.



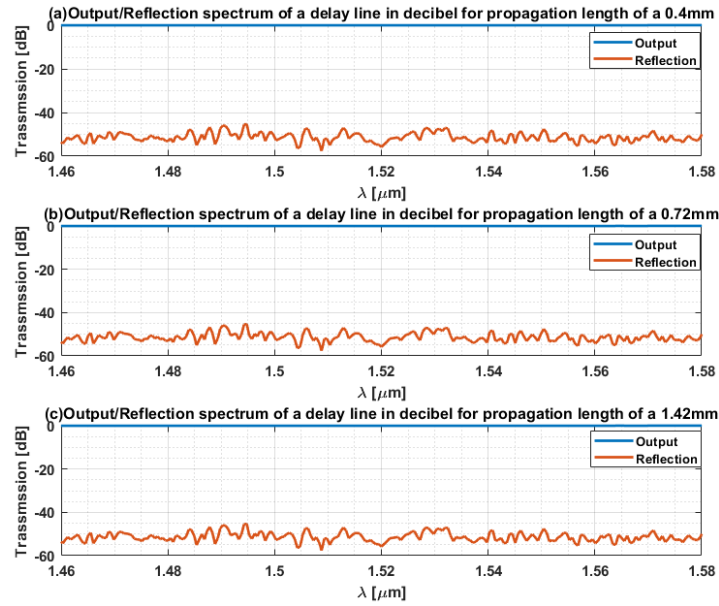
63
64
65
66
67
68

Fig.1. Schematic of the Archimedean true optical delay line (TODL) consists of two interleaved waveguides with gradually decreasing bending radiuses. As they reach the center of the spiral, those two spirals are connected in the center by an S-shape bend structure. Clockwise direction (red) is the input waveguide, and (blue) is the output waveguide, where both are separated by a gap(g).

69 Lumerical 2.5D variational finite-difference time-domain (varFDTD) simulations are used
70 to investigate the effect of the width and gap of waveguides 1 and 2 on the performance of the
71 TODL, as shown in Fig. 1. Specifically, the performance of three design cases was investigated
72 and compared: in design (1) both waveguides were identical where the width was set to be 550
73 nm with a 1 μm gap, in design (2) both waveguides were identical where the width was set to
74 be 550nm with a 300 nm gap, and in design (3) the width of waveguide 1 was 550 nm while
75 the width of waveguide 2 was 400 nm with a 300 nm gap. Two monitors were setup in each of
76 these simulations; one monitor was placed at the end of the spiral to measure the output
77 transmission spectrum, denoted as output. The second monitor is located behind the input
78 source, as seen in Fig.1, to measure the reflected spectrum back to the input, denoted as
79 reflection. These monitors measure the reflected and transmitted spectrum that results from
80 signal propagating through the optical delay line (TODL). The reflected spectrum is an
81 indication of the strength of the interaction between the adjacent waveguides in the spiral delay

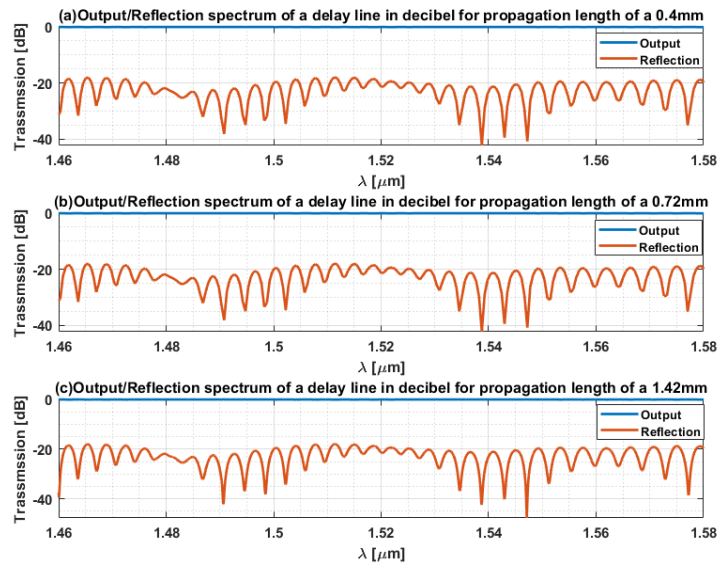
82 line. This interaction causes some transmitted spectrum to propagate back to the input direction.
 83 Typically, a smaller reflected spectrum results from minimal interaction between the
 84 waveguides in the TODL. Furthermore, a larger reflected spectrum is caused by high interaction
 85 between the waveguides as the mode propagates through the spiral.

86 We expect to see a very low to non-existing reflected transmission spectrum from the design
 87 simulation (1) due to a large separation gap of $1\mu\text{m}$, which is the typical separation gap used
 88 for TODL in literature [16,19]. In the design simulation (2), the reflected spectrum is due to a
 89 sub-micron gap and has no geometric dispersion since waveguides 1 and 2 have the exact
 90 dimensions. However, the effect of bending dispersion can cause variation in the propagation
 91 constant (β) and thus might reduce the total reflected spectrum. Finally, the design simulation
 92 (3) has both geometry dispersion and bending dispersion, which offers a unique opportunity to
 93 observe both the effect on the propagation constant (β) and reflected mode strength.
 94



95
 96
 97
 98
 99
 100
 101
 102
 Fig. 2. Transmission spectrums of the simulated Archimedes spiral delay line for different propagation lengths where waveguides 1 and 2 are identical (i.e., width = 550 nm) and the gap between them is $1\mu\text{m}$. The blue indicates the transmitted spectrum through the suggested spiral, while the orange indicates the reflected spectrum back to the source. (a) Transmission spectrum of a delay line in decibel for propagation length of 0.4 mm. (b) Transmission spectrum of a delay line in decibel for propagation length of 0.72 mm. (c) Transmission spectrum of a delay line in decibel for propagation length of a 1.42 mm.

103 Fig. 2 shows the simulation of design (1), where waveguides 1 and 2 have a similar width
 104 of 550 nm and a separating gap of $1\mu\text{m}$. The figure shows the results of transmission and
 105 reflection spectrums of an Archimedes delay line in the spectral band of 1.46-1.58 μm when its
 106 adjacent waveguides are identical (i.e., 550 nm) with a $1\mu\text{m}$ gap for a total length of 0.4, 0.72,
 107 and 1.42 mm as shown in Figs 2 (a-c). The results in Fig. 2 show the transmitted spectrum in
 108 decibel to be 0 dB while the reflected signal is maintained at less than -40 dB for all propagation
 109 lengths of 0.4, 0.72, and 1.42 mm. It can be concluded that, as expected, there is no interaction
 110 between the adjacent waveguide in the Archimedes spiral. Also, the reflected spectrum does
 111 not vary as a function of propagation length which confirms that there is no coupling in this
 112 Archimedes spiral TODL. It is important to note that even though the waveguides have the
 113 same width, they experience different propagation constant (β) due to the bending dispersion
 114 due to the low bending radius [21]. This effect is not visible here due to the large separation
 115 gap in this design; however, this dispersion effect is observed in Fig.3 and Fig.4.



116
117
118
119
120
121
122
123

Fig. 3. Transmission spectrum of the simulated Archimedes spiral delay line with different propagation lengths where both waveguide widths are identical (i.e., width = 550 nm) and the gap between them is 300 nm. The blue indicates the transmitted spectrum through the suggested spiral, while the orange indicates the reflected spectrum back to the source. (a) Transmission spectrum of a delay line in decibel for propagation length of 0.4mm. (b) Transmission spectrum of a delay line in decibel for propagation length of 0.72mm. (c) Transmission spectrum of a delay line in decibel for propagation length of 1.42mm.

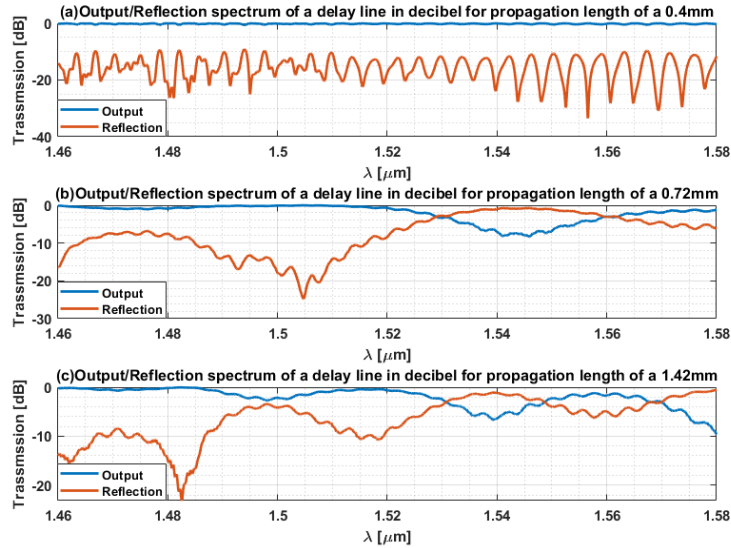
124
125
126
127
128
129
130
131
132
133
134
135
136
137
138
139

Fig. 3 shows the simulation of design (2), where waveguides 1 and 2 have similar widths of 550 nm with a separating gap of 300 nm. An oscillation of less than 0.1% is observed in the transmitted spectrum for all propagation lengths of 0.4, 0.72, and 1.42 mm (see Fig.3(a-c)). A comparison of the results of simulated design (1) in Fig. 2 and simulated design (2) in Fig. 3 shows that the reflected spectrum increases from approximately -40dB to -20dB when the gap is decreased from 1 μm to 300nm. The reflected spectrum does not vary as a function of the length of the delay line, which shows that the transmission spectrum is not significantly affected by any resonant phenomenon caused by the interaction between the adjacent waveguides under a small separation gap. If such resonance were to occur, there would be a clear dependence of the reflected spectra on the spiral length (because each length would have a different resonant cavity length). Instead, the signal-to-noise ratio increase is more likely a result of imperfect simulation meshing of the curved waveguide. The result indicates a negligibly small interaction between the two adjacent waveguides. Although this design is not affected by geometry dispersion, it is effect by the bending dispersion, which causes a different propagation constant(β) for each waveguide; such a difference in the propagation constant limits the interaction in the spiral.

140
141
142
143
144
145
146
147
148
149

Fig. 4 shows the simulation of design (3), where the width of waveguide 1 was 550 nm while the width of waveguide 2 was 400 nm with a 300 nm gap. For the delay line of 0.4 mm length, we observe an approximate 5% oscillation in the transmission spectrum (see Fig.4(a)), significantly increasing across the spectrum for the delay lengths of 0.72mm (see Fig. 4(b)) and 1.42mm (see Fig.4(c)). This oscillatory behavior can be characterized as a result of the formation of a distributed spiral resonator due to coupling between adjacent waveguides, where its cavity length varies with the total length of the spiral optical delay line. With this hypothesis, the Free Spectral Range (FSR) for the 0.72 mm and 1.42 mm long spirals in Fig.4(b) and Fig. 4(c) is approximately 66.7 nm and 50nm, respectively. The corresponding Q factors for these resonators are estimated to be 38.8 and 50 for spiral lengths of 0.72 and 1.42 mm, respectively.

150 When comparing Fig. 3(b) and Fig. 4(b), we observe that when the adjacent waveguides are
 151 identical ($\Delta w=0$), the reflected signal is at -20 dB, while when $\Delta w=150$ nm, the reflected signal
 152 oscillates between -10 to 0 dB. We can conclude that in this design, a resonance effect is caused
 153 by coupling between the adjacent waveguides. This coupling is enhanced by the combined
 154 effect of the geometry and bending dispersion. However, the observed resonator needed to be
 155 analyzed further to understand it.



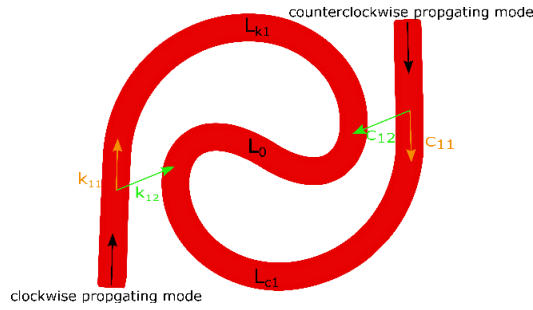
156
 157 Fig. 4. Transmission spectrum of the simulated Archimedes spiral delay line with different
 158 propagation lengths where waveguide 1 and 2 widths are 550 and 400nm, respectively, and the
 159 gap between them is 300 nm. The blue indicates the transmitted spectrum through the suggested
 160 spiral, while the orange indicates the reflected spectrum back to the source. (a) Transmission
 161 spectrum of a delay line in decibel for propagation length of 0.4mm. (b) Transmission spectrum
 162 of a delay line in decibel for propagation length of 0.72mm. (c) Transmission spectrum of a
 163 delay line in decibel for propagation length of 1.42mm.

164 3. Analysis of the resonant behavior in the Archimedes TODL

165 We next discuss a simplified analytical model to describe the resonant transmission and
 166 reflection of a closely packed Archimedes spiral TODL. When formulating the problem
 167 mathematically, it is tempting to try the matrix approach that would solve the problem in terms
 168 of supermodels. However, this approach runs into problems due to the spatial variation of the
 169 coupling coefficient along the spiral length.

170 Consequently, it is desirable to use a more flexible approach that uses reflection coefficients to
 171 represent the coupling regions in the spiral. The repetitive structure of the spiral lends well to
 172 this analysis approach.

173 As an initial approach, we consider a simplified spiral version, as seen in Fig.5, Where the
 174 coupling is limited to a small section of each arm in the S-shape bend. The field coupling
 175 coefficients k and c for the clockwise and counterclockwise propagation with the subscripts
 176 indicating the coupling direction.



177

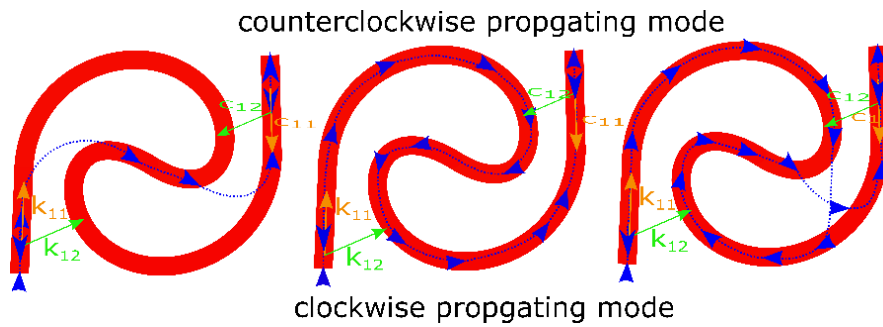
178

Fig. 5. Simplified coupling model in a resonator formed by a single loop spiral.

179

Consider the transmission coefficient as a summation similar to Airy's formulas (multiple resonant paths) illustrated in Fig. 6 for the Fabry–Pérot resonator:

180



181

182

Fig. 6. The possible path the propagated mode can take.

$$\begin{aligned}
 t_s = & \underbrace{k_{12} \exp(i\beta L_0) c_{12}}_{\text{nonrepeating}} \\
 & + \underbrace{k_{12} \exp(i\beta L_0) c_{11} \exp(i\beta L_{k_1}) k_{12} \exp(i\beta L_{c_1})}_{\text{beginning 1}} \underbrace{c_{11}}_{\text{ending 1}} \\
 & + \underbrace{k_{11} \exp(i\beta L_{k_1}) c_{11} \exp(i\beta L_0) k_{11} \exp(i\beta L_{c_1})}_{\text{beginning 2}} \underbrace{c_{11}}_{\text{ending 1}} \\
 & + \underbrace{k_{11} \exp(i\beta L_{k_1}) c_{12} \exp(i\beta L_{c_1}) k_{11} \exp(i\beta L_0) c_{12}}_{\text{beginning 3}} \underbrace{c_{12}}_{\text{ending 2}} \\
 & + \underbrace{k_{12} \exp(i\beta L_0) c_{11} \exp(i\beta L_{k_1}) k_{12} \exp(i\beta L_{c_1}) c_{12} \exp(i\beta L_{k_1}) k_{12} \exp(i\beta L_{c_1})}_{\text{beginning 1}} \underbrace{c_{11}}_{\text{ending 1}} \\
 & + \underbrace{k_{11} \exp(i\beta L_{k_1}) c_{11} \exp(i\beta L_0) k_{11} \exp(i\beta L_{c_1}) c_{12} \exp(i\beta L_{k_1}) k_{12} \exp(i\beta L_{c_1})}_{\text{beginning 2}} \underbrace{c_{11}}_{\text{ending 1}} \\
 & + \underbrace{k_{11} \exp(i\beta L_{k_1}) c_{12} \exp(i\beta L_{c_1}) k_{12} \exp(i\beta L_{k_1}) c_{12} \exp(i\beta L_{c_1}) k_{11} \exp(i\beta L_0) c_{12}}_{\text{beginning 3}} \underbrace{c_{12}}_{\text{ending 2}} \\
 & + \dots
 \end{aligned}$$

183

184

Although the terms of the summation in the simplest spiral are much more complex than those in a Fabry–Pérot cavity [22], they simplify similarly. Reduce the above terms using a geometric series as follows:

185

186

187

$$\sum_{k=0}^{\infty} ar^k = \frac{a}{1-r}, r < 1$$

188

189
$$t_s = k_{12} \exp(i\beta L_0) c_{12} + (k_{12} c_{11} k_{12} c_{11} + k_{11} c_{11} k_{11} c_{11} + k_{11} c_{12} k_{11} c_{12}) \frac{\exp(i\beta L_0) \exp(i\beta L_{k1}) \exp(i\beta L_{c1})}{1 - c_{12} \exp(i\beta L_{k1}) k_{12} \exp(i\beta L_{c1})}$$

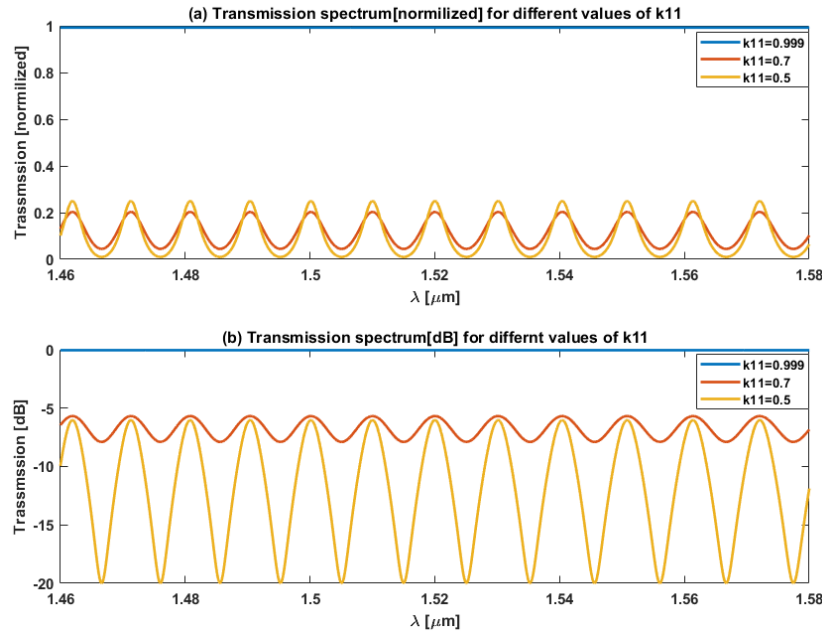
190 Moreover, the reflection coefficient can be written as:

191
$$r_s = \left[1 + k_{11}^2 \exp(i\beta L_{c1}) \exp(i\beta L_{k1}) \right] \frac{2k_{11} k_{12} c_{11} \exp(i\beta L_0) \exp(i\beta L_{k1})}{1 - k_{12} \exp(i\beta L_{c1}) c_{12} \exp(i\beta L_{k1})}$$

192 These expressions are similar to the Fabry–Pérot reflection coefficient equation. The leading
 193 constant terms make analytical evaluation difficult; however, they can be calculated
 194 numerically. For the simplest case of the Archimedes delay line illustrated in Fig. 5, we estimate
 195 47.5, 24.8, and 47.5 μm corresponding to Lc1, L0, and Lk1's cavity lengths, respectively. To
 196 further simplify the equation above, the two coupling terms k_{11} and c_{11} can be related thus:

197
$$c_{11} = 1 - k_{11}$$

198 In addition, assuming that the two sides of the simplified version of the distributed spiral
 199 resonator are similar, it is safe to assume that $c_{11} = c_{12}$ and $k_{11} = k_{12}$. With these values, it is
 200 possible to plot the analytic solution from the transmitted spectrum of the equation above as a
 201 function of the coupling coefficient k_{11} , shown in Fig. 7.



202

203 Fig. 7. Transmission spectrum as a function of cross coupling parameter k_{11} for ideal resonator
 204 loop where $|r|^2 + |t|^2 = 1$. (a) normalized transmission spectrum for k_{11} [blue=0.999, red=0.7,
 205 yellow=0.5]. (b) Transmission spectrum in decibel for k_{11} [blue=0.999, red=0.7, yellow=0.5]

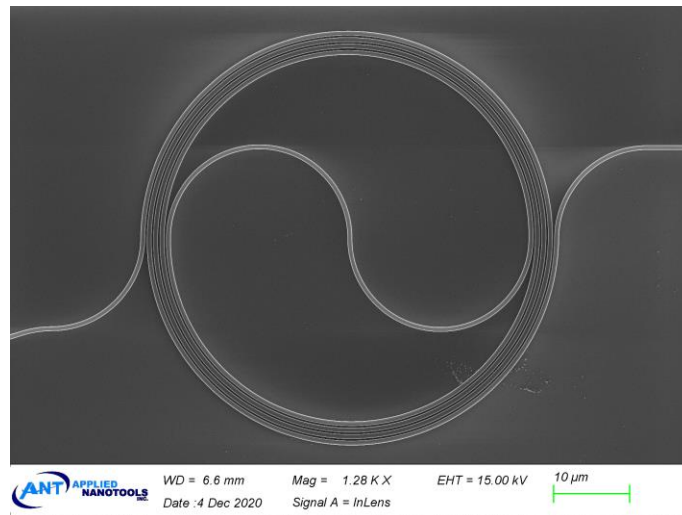
206 Fig. 7 (a[blue line]) shows the transmission spectrum of the suggested model, in which the
 207 coupling coefficient (k_{11}) is equal to 0.999. The same resonance is observed in the transmission
 208 spectrum in Fig. 2. Fig. 7(a[red line]) shows the effect of the distributed Fabry–Pérot-like
 209 resonator when the coupling coefficient $k_{11} = 0.7$, which has an FWHM equal to 3.2 nm
 210 corresponding to a Q factor of 458.3. Moreover, Fig. 7 (a[yellow line]) shows the same effect
 211 when the coupling coefficient $k_{11} = 0.5$, which has an FWHM equal to 5.8 nm corresponding to
 212 a Q factor of 252. By changing the coupling coefficient (k_{11}), the effect of the distributed spiral

213 resonator Q factor can be reduced. For this manuscript, the Fabry–Pérot-like resonator effect is
214 undesirable and can be reduced by limiting the cross-coupling coefficient. Conversely, for other
215 applications, a distributed spiral resonator might be helpful. In such a case, the primary way
216 this can be achieved, as seen from the simulations in Fig.6 and Fig.5, is not to introduce a
217 variation between the propagation constants of the two adjacent waveguides.

218 Strictly speaking, the above process could be used to analyze a spiral with multiple turns.
219 However, the sheer number of new closed-loop paths introduced by each additional turn makes
220 this quickly become resistant to analysis. However, the model is still helpful as it provides
221 enough information that we can develop the intuition that these additional path loops will
222 behave additively in the same way as the loops in the simple model. Therefore, we know how
223 the spiral will behave in a general sense (there will be distinct resonance peaks) even if it is
224 difficult to predict the exact spacing in practice accurately.

225 4. Fabrication and characterization

226 The Archimedes spiral TODL design was fabricated as part of a multi-project-wafer (MPW)
227 run at the Applied Nanotools (ANT) foundry. The MPW run utilizes standard silicon on
228 insulator (SOI) wafer, which consists of a 220 nm device layer and a 2 μm bottom oxide layer
229 (BOX). The standard process of the MPW run starts with writing the design pattern on e-beam
230 resist using electron beam lithography (EBL). Inductively Coupled Plasma Etching (ICP-RIE)
231 is employed to transfer the pattern from the E-beam resist into the silicon layer. The sample is
232 cladded with a 2.2 μm SiO_2 layer through Plasma Enhanced Chemical Vapor Deposition
233 (PECVD). The overall design consists of multiple Archimedes spirals with different gaps,
234 widths, and several spirals to demonstrate the simulation results, as seen in Fig.8.



235

236

237

238

Fig.8. The SEM images of the fabricated Archimedes spiral delay lines: (a) The input and the output waveguides have similar dimensions of 550nm width, 220nm height, and a separation gap of 300nm with a total length of 0.72mm.

239

240

241

The device was characterized using the standard fiber-to-free space setup described previously in [23,24]. Finally, the propagation loss of the fabricated Archimedes spiral delay line is specified to be 3.8 dB/cm [25,26].

242

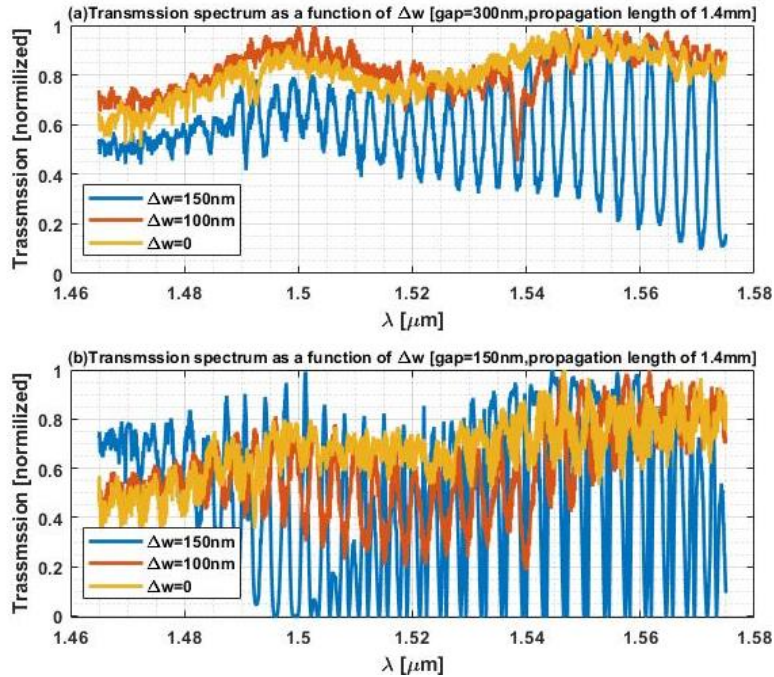
5. Results

243

244

245

The fabricated Archimedes spiral of the delay line is designed to observe the effect of introducing a width difference (Δw) of 0, 100, or 150 nm for the cases of a separation gap of 300nm and 150nm in Fig.9.



246

247
248
249
250
251
252
253
254

Fig.9. Measured transmission spectrum of the fabricated Archimedes spiral delay line with three different Δw at a total length of 1.4mm. The blue color indicates a difference between the adjacent waveguide (Δw) equals 150 nm. The Red color indicates a difference between the adjacent waveguide is (Δw) equal to 100 nm; the yellow color shows the difference between the adjacent waveguide is (Δw) zero. (a) measured transmission spectrum of the fabricated delay line propagation length equal to 1.4mm and separation gap equal to 300nm. (b) measured transmission spectrum of the fabricated delay line propagation length equal to 1.4mm and separation gap equal to 150nm.

255
256
257
258
259
260
261

Fig.9(a) shows the waveguide width difference (Δw) effect on the transmission spectrum at a gap of 300 nm and spiral length of 1.4 mm. In the transmission spectrum, when $\Delta w=0$ nm, no resonance exists in the range from 1460 nm to 1540 nm; a slight variation can be noticed from above the wavelength of 1540 nm with a variation of less than 10%. Thus, we can conclude that bending dispersion minimizes a distributed spiral resonator effect. In addition, the transmission spectrum variation percentage when $\Delta w=0$ and 100 nm is less than 9% and 20%, respectively.

262
263
264
265
266
267
268
269

However, when $\Delta w=150$ nm, the transmitted spectrum varies from 30% for wavelengths between 1460 nm to 1480 nm, to 40%, for wavelengths between 1500nm to 1540nm, in which the maximum transmission spectrum variation is shown to be 78% for wavelengths between 1560nm to 1580nm. The transmission spectrum variation shows a clear wavelength dependence, resulting from geometry dispersion-compensating for the variation in propagation constant (β) caused by bending dispersion. This effect increases the strength of coupling between the adjacent waveguides in the spirals and thus enhancing the distributed spiral resonator. The same enhancement is noticeable when $\Delta w=100$ nm, the average transmission variation is less than 20%, and the maximum variation in the transmission spectrum is 35%.

270
271
272
273
274
275
276

Fig.9(b) shows the transmission spectrum of an Archimedes delay line with a separation gap equal to 150 nm. At this gap, the effect of the distributed spiral resonator can be observed in all different Δw values. When $\Delta w=150$ nm, the transmission varies from 40% between 1460 and 1480 nm to 84% for wavelengths between 1500 and 1540 nm. Moreover, the transmission variation is 87% for wavelengths between 1560 to 1580 nm. The same effect is noticeable when $\Delta w=100$ nm with a minor transmission variation compared to when $\Delta w=150$ nm. As described

277 in the literature, coupling between waveguides is a function of the separation gap [20], hence
 278 the increased cross-coupling from the change gap from 300 to 150 nm. Furthermore, the
 279 coupling coefficient's wavelength dependence is a clear indication that the coupling which
 280 causes the distributed spiral resonator is enhanced by the smaller gap, especially when
 281 comparing these results to Fig.10(a). The presence of high resonance effects suggests that the
 282 gap of 150nm cannot be used for a typical delay line due to the high effect of the distributed
 283 spiral resonator.

284 It is important to note that the reflected spectrum is not measured due to the limitation of
 285 the current fiber-to-free space setup and the fabricated sample. In the future, a sample can be
 286 fabricated to include a Y-splitter at the input to extract the reflected spectrum easily.

287 6. Discussion

288 In this work, we showed a new approach to maximizing the packing density of TODL by
 289 reducing the separation gap to submicron and maintaining a minimum bending radius while
 290 also avoiding inducing radial loss. As observed in the simulation, these conditions give rise to
 291 a novel type of distributed spiral resonator. This resonator effect can be mitigated by
 292 introducing enough variation in the preparation constant (β) between the adjacent waveguides.

293 While the bending radius is maintained to be small enough, the two adjacent waveguides
 294 are kept in the exact dimensions to minimize coupling and thus eliminate the effect of the
 295 distributed spiral resonator. Moreover, the variation caused by the propagation constant (β) will
 296 become negligible in the case of a large enough bending radius. In that case, it would be
 297 necessary to vary the cross-section of adjacent waveguides to achieve geometry dispersion, thus
 298 introducing a mismatch in the propagation constant (β).

299 The presented results validate the suggested theoretical model and simulated results in Fig.3
 300 and Fig.4. The effect of the distributed spiral resonator can be enhanced when the two
 301 waveguides have similar propagation constants, as in the case of Fig.10 when $\Delta w=150$ nm.
 302 Additionally, the effect of the distributed spiral resonator can be reduced when the waveguides
 303 have different propagation constants, as in the case of Fig.10 when $\Delta w=0$.

304 The fabricated design in the manuscript is compared to some of the delay line structures
 305 found in the literature (see Table1). The last column in Table 1 shows a new figure of merit
 306 called Linear Density Figure of Merit (LDFM), which evaluates the packing efficiency of
 307 various delay line design approaches.

308 **Table 1. Comparison of various optical delay lines from literature and the fabricated design**

Paper	Loss	Time delay(τ)	Approach	Bandwidth	Material	LDFM (km^{-1})
Bykhovsky [27]	0.5 dB	17.2ns	Cascaded spiral	80nm	Si_3N_4	146
Yan Li [28]	1.908 dB/cm	2.804ns	Archimedes spiral	NA	SOI	197.9
Chen [29]	0.1 dB/m	NA	Whispering gallery delay lines	NA	Si/ SiO_2	77.56
Yurtsever [30]	<0.14 dB/cm	NA	Mach-Zehnder interferometer	100nm	$\text{Si}_3\text{N}_4/\text{SiO}_2$	0.575
Stopinski [31]	10 dB	250 ps	Archimedes spiral	30 GHz	InP	52.632
This work	3.8 dB/cm	19.41 ps	Archimedes spiral	100nm	SOI	388

310 LDFM is an excellent method to evaluate the cost-effectiveness of any TODL since it
311 occupies a large area of the chip to achieve a long delay time (τ). A more significant LDFM
312 number indicates a higher packing density and thus a more cost-effective method to achieve
313 the required delay time. This aspect of delay line design is often overlooked but very important
314 to classify because the available space for integrated devices is often minimal. Finally, The
315 proposed design in this paper shows the highest LDFM with a value of 388 km^{-1} , which is at
316 least two times higher than the designs reported in recently published papers, as seen in Table 1.
317 The loss of the fabricated TODL is 3.8 dB/cm , which can improve by choosing a wider
318 waveguide or changing to material that exhibits lower loss (for example, Si_3N_4), as
319 demonstrated in Bykhovsky's work [27].

320 7. Conclusion:

321 In conclusion, we have designed, fabricated, and characterized a broadband Archimedes spiral
322 delay line with a high packing density. The higher packing density is achieved by reducing the
323 gap between two adjacent waveguides in the sub-micron range and maintaining a low bending
324 radius. Waveguides with such slight separation typically experience substantial coupling-
325 induced crosstalk that limits the device's spectral response. It was experimentally demonstrated
326 that this effect could be mitigated by engineering the waveguides in different arms of the spiral
327 to have different propagation constants. Furthermore, to validate this design approach, we
328 developed an analytical model of the spiral that explicitly incorporates the evanescent coupling.
329 This model shows that when the coupling is non-negligible, the device functions as a novel
330 type of distributed spiral resonator. It was concluded that while maintaining a low bending
331 radius, each waveguide experiences a different propagation constant due to the bending
332 dispersion; thus, maintaining similar geometry for the waveguide in the spiral would limit the
333 distributed spiral resonator effect. Finally, an LDFM is defined to evaluate the packing
334 efficiency and cost-effectiveness of future delay line design approaches.
335

336 **Funding.** Defense Advanced Research Projects Agency (DSO NAC Programs); Office of Naval Research; National
337 Science Foundation (ECCS-180789, NSF ECCS-190184, NSF ECCS-2023730); Army Research Office; San Diego
338 Nanotechnology Infrastructure (SDNI) supported by the NSF National Nanotechnology Coordinated Infrastructure
339 (ECCS-2025752); Quantum Materials for Energy Efficient Neuromorphic Computing-an Energy Frontier Research
340 Center funded by the U.S. Department of Energy (DOE) Office of Science; Basic Energy Sciences (DE-SC0019273);
341 LEED: A Lightwave Energy-Efficient Datacenter funded by the Advanced Research Projects Agency-Energy; Cymer
342 Corporation.

343 **Acknowledgments.** Dhaifallah Almutairi and Naif Alshamrani would like to thank King Abdulaziz City for Science
344 and Technology (KACST) for their support during their study.

345 **Disclosures.** The authors declare no conflicts of interest.

346 **Data availability.** Data underlying the results presented in this paper are not publicly available at this time but
347 may be obtained from the authors upon reasonable request.

348 Reference:

- 349 1. R. W. Boyd, D. J. Gauthier, and A. L. Gaeta, "Applications of slow light in telecommunications," *Opt. Photonics*
350 *News* 17, 18–23 (2006)
- 351 2. A. E. Willner, B. Zhang, L. Zhang, L. Yan, and I. Fazal, "Optical signal processing using tunable delay elements
352 based on slow light," *IEEE J. Sel. Top. Quantum Electron.* 14, 691–705 (2008).
- 353 3. J. Capmany, B. Ortega, and D. Pastor, "A tutorial on microwave photonic filters," *J. Lightwave Technol.* 24,
354 201–229 (2006).
- 355 4. C. Ciminelli, C. E. Campanella, F. Dell'Olivo, M. N. Armenise, E. Armandillo, and I. McKenzie, "Study of
356 photonic resonant angular velocity sensors as alternative gyro technology," *Proc. SPIE* 10564, 105641M (2017).
- 357 5. G. Yurtsever, B. Považay, A. Alex, B. Zabihian, W. Drexler, and R. Baets, "Photonic integrated Mach-Zehnder
358 interferometer with an on-chip reference arm for optical coherence tomography," *Biomed. Opt. Express* 5, 1050
359 (2014).
- 360 6. E. N. Toughlian and H. Zmuda, "A photonic variable RF delay line for phased array antennas," in *Journal of*
361 *Lightwave Technology*, vol. 8, no. 12, pp. 1824–1828, Dec. 1990, doi: 10.1109/50.62877.

362 7. Lvovsky, A., Sanders, B. & Tittel, W. Optical quantum memory. *Nature Photon* 3, 706–714 (2009).
363 <https://doi.org/10.1038/nphoton.2009.231>.

364 8. Vandoorne, K., Mechet, P., Van Vaerenbergh, T. et al. Experimental demonstration of reservoir computing on a
365 silicon photonics chip. *Nat Commun* 5, 3541 (2014). <https://doi.org/10.1038/ncomms4541>.

366 9. Jahn, D., Lippert, S., Bisi, M. et al. On the Influence of Delay Line Uncertainty in THz Time-Domain
367 Spectroscopy. *J Infrared MilliTerahz Waves* 37, 605–613 (2016). <https://doi.org/10.1007/s10762-016-0250-4>.

368 10. I. Giuntoni, D. Stolarek, D. I. Kroushkov, J. Bruns, L. Zimmermann, B. Tillack, and K. Petermann,
369 "Continuously tunable delay line based on SOI tapered Bragg gratings," *Opt. Express* 20, 11241 (2012).

370 11. Y. A. Vlasov, M. O'Boyle, H. F. Hamann, and S. J. McNab, "Active control of slow light on a chip with photonic
371 crystal waveguides," *Nature* 438, 65–69 (2005).

372 12. D. O'Brien, A. Gomez-Iglesias, M. D. Settle, A. Michaeli, M. Salib, and T. F. Krauss, "Tunable optical delay
373 using photonic crystal heterostructure nanocavities," *Phys. Rev. B* 76, 115110 (2007).

374 13. Joyce K. S. Poon, Jacob Scheuer, Yong Xu, and Amnon Yariv, "Designing coupled-resonator optical waveguide
375 delay lines," *J. Opt. Soc. Am. B* 21, 1665-1673 (2004)

376 14. Sanghoon Chin, Luc Thévenaz, Juan Sancho, Salvador Sales, José Capmany, Perrine Berger,
377 Jérôme Bourderionnet, and Daniel Dolfi, "Broadband true time delay for microwave signal processing, using slow
378 light based on stimulated Brillouin scattering in optical fibers," *Opt. Express* 18, 22599-22613 (2010) 15. D.

379 15. Marpaung, B. Morrison, M. Pagani, R. Pant, D.-Y. Choi, B. Luther-Davies, S. J. Madden, and B. J. Eggleton,
380 "Low-power, chip-based stimulated Brillouin scattering microwave photonic filter with ultrahigh selectivity,"
381 *Optica* 2, 76 (2015).

382 16. Mégret, P. et al. "Ring resonator-based Tunable Optical Delay Line in LPCVD Waveguide Technology." (2005).

383 17. Guoliang Li, Jin Yao, Hiren Thacker, Attila Mekis, Xuezhe Zheng, Ivan Shubin, Ying Luo, Jin-hyoung Lee,
384 Kannan Raj, John E. Cunningham, and Ashok V. Krishnamoorthy, "Ultralow-loss, high-density SOI optical
385 waveguide routing for macrochip interconnects," *Opt. Express* 20, 12035-12039 (2012).

386 18. Lee, H., Chen, T., Li, J. et al. Ultra-low-loss optical delay line on a silicon chip. *Nat Commun* 3, 867 (2012).
387 <https://doi.org/10.1038/ncomms1876>.

388 19. X. Ji, X. Yao, Y. Gan, A. Mohanty, M. A. Tadayon, C. P. Hendon, and M. Lipson, "On-chip tunable photonic
389 delay line," *APL Photonics* 4(9), 090803 (2019).

390 20. Yariv, A., & Yeh, P. (2003). chapter 11. In *Optical waves in crystals: Propagation and control of Laser Radiation*
391 (pp. 425–475). essay, John Wiley and Sons.

392 21. M. Heiblum and J. Harris, "Analysis of curved optical waveguides by conformal transformation," in *IEEE Journal of*
393 *Quantum Electronics*, vol. 11, no. 2, pp. 75-83, February 1975, doi: 10.1109/JQE.1975.1068563.

394 22. Vaughan, John Michael. *The Fabry–Perot interferometer: history, theory, practice and applications*. Routledge,
395 2017.

396 23. N. Alshamrani, A. Grieco, B. Hong, and Y. Fainman, "Miniaturized integrated spectrometer using a silicon ring-
397 grating design," *Opt. Express* 29(10), 15279–15287 (2021).

398 24. A. Grieco, B. Slutsky, D. T. H. Tan, S. Zamek, M. P. Nezhad, and Y. Fainman, "Optical bistability in a silicon
399 waveguide distributed bragg reflector Fabry-pérot resonator," *J. Lightwave Technol.* 30(14), 2352–2355 (2012)

400 25. Appliednt.com. 2020. Nanosoi Fabrication Service | Applied NanotoolsInc.. [online] Available at:
401 <https://www.appliednt.com/nanosoi-fabrication-service/>

402 26. Appliednt.com. 2020. Nanosoi Fabrication Service | Applied NanotoolsInc.. [online] Available at:
403 <https://www.appliednt.com/nanosoi/sys/resources/specs/>.

404 27. Bykhovsky, D., Rosenblit, M., & Arnon, S. (2018). Two-sided through-wafer interconnect for optical spiral delay
405 line. *Journal of Modern Optics*, 65(1), 98–103. <https://doi.org/10.1080/09500340.2017.1377305> Y. Li, X. Song,
406 Y. Xu, and Y. Li, "Spiral Optical Delay Lines in Silicon-on-Insulator," in *Asia Communications and Photonics*
407 *Conference 2016, OSA Technical Digest (online)* (Optical Society of America, 2016), paper AF3G.6.

408 28. JingyaXie, Linjie Zhou, Zuxiang Li, Jinting Wang, and Jianping Chen, "Seven-bit reconfigurable optical true
409 time delay line based on silicon integration," *Opt. Express* 22, 22707-22715 (2014)

410 29. Chen, T., Lee, H., & Vahala, K. J. (2014). Design and characterization of whispering-gallery spiral waveguides.
411 *Optics Express*, 22(5), 5196. <https://doi.org/10.1364/oe.22.005196>

412 30. Yurtsever, G., Považay, B., Alex, A., Zabihiyan, B., Drexler, W., & Baets, R. (2014). Photonic integrated Mach-
413 Zehnder interferometer with an on-chip reference arm for optical coherence tomography. *Biomedical Optics*
414 *Express*, 5(4), 1050. <https://doi.org/10.1364/boe.5.001050>.

415 31. Stopinski, S., Malinowski, M., Piramidowicz, R., Kleijn, E., Smit, M. K., & Leijtens, X. J. M. (2013). Integrated
416 optical delay lines for time-division multiplexers. *IEEE Photonics Journal*,
417 5(5). <https://doi.org/10.1109/JPHOT.2013.2280519>.

418

419